

Low Noise Avalanche Photodiodes

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Talk Outline

- APD background
- **♦ Low noise mechanisms in thin p⁺-i-n⁺s**
- **→** Temperature dependence
- → APD speed
- Conclusions

National Centre for III-V Technologies, University of Sheffield, UK



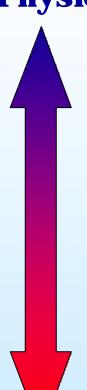


- ★ Established in Department of Electronic & Electrical Engineering, University of Sheffield in 1978
- ★ Mission: To provide III-V wafers & devices to the UK academic community
- **★ Current Capability:** 2 MBE, 3 MOVPE, Device Fabrication, Characterisation
- * Staff: 10 scientists, 6 technicians
- ★ **Growth output:** 750 wafers/year
- ★ Optical wafers & devices: Lasers, LEDs, VCSELs, RC-LEDs, waveguides, modulators, AFPMs, pins, APDs, Q-Dot lasers, Q-Cascade lasers
- ★ Electrical wafers & devices: HBTs, HEMTs, diodes

APD Research at Sheffield



Physics

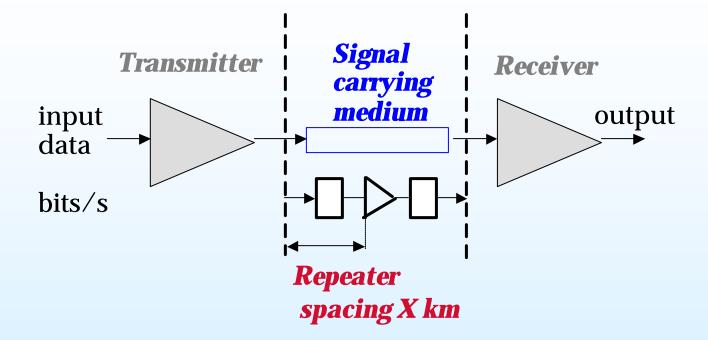


- Impact ionization coefficient investigation
- New materials and novel structures
- 'Dead-space' characterization
- Excess noise measurements
- Temperature dependence
- Analytical and numerical modelling
- Low-noise, high-speed avalanche photodiodes
- Single photon avalanche photodiodes (SPADs)

Devices

Generic communication system



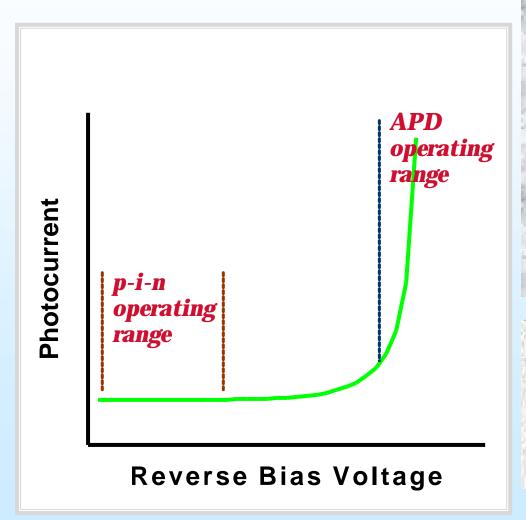


Communication capacity - Bit rate length product = repeater spacing*data rate {X [km]*Bandwidth [b/s]}

Using an APD can increase repeater spacing

p-i-n vs. APD





p-i-n

- * low voltage
- * bias insensitive
- * temperature insensitive
- * simple bias circuit
- * fast
- * simple bias circuit
- * cheap

APD

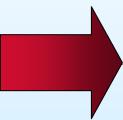
- * high sensitivity
- * single photon detection (possibility)

Technology Comparison



Photomultipliers

- + High gain (~10⁶)
- + Low dark current
- + Low noise
- ? Reliability
- Expensive
- Poor efficiency
- Complex filters
- Bulky
- Fragile
- High voltage (~1kV)



Avalanche photodiodes

- + Inexpensive
- + Compact
- + Rugged
- + High detectivity
- + High reliability
- + Simpler, cheaper filters
- + Reasonable gain
- + High efficiency
- + Low voltage (<100V)
- High noise
- Temperature sensitive

Comparison of three detector types

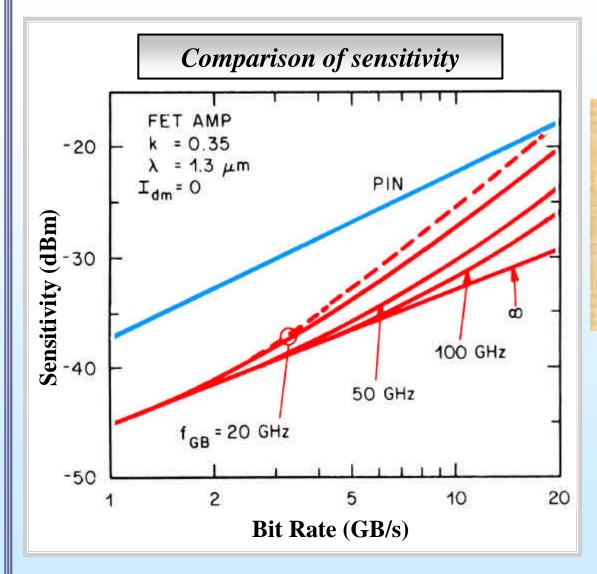


	PIN-FET	Ge APD	InGaAs APD
Sensitivity	X dBm	(X+4) dBm	(X+8) dBm
Cost	Moderate	Moderate	High
Wavelength	1.3 _m m & 1.5 _m m	1.3 _m m	1.3 _m m & 1.5 _m m
Reliability	10 ¹¹ hrs	10 ⁶ hrs	10 ⁶ hrs

- ◆ An InGaAs APD requires 8dB less optical power to produce the same signal as a PIN-FET.
- ◆ PIN-FET is cheaper, faster and more reliable.

p-i-n vs. APD

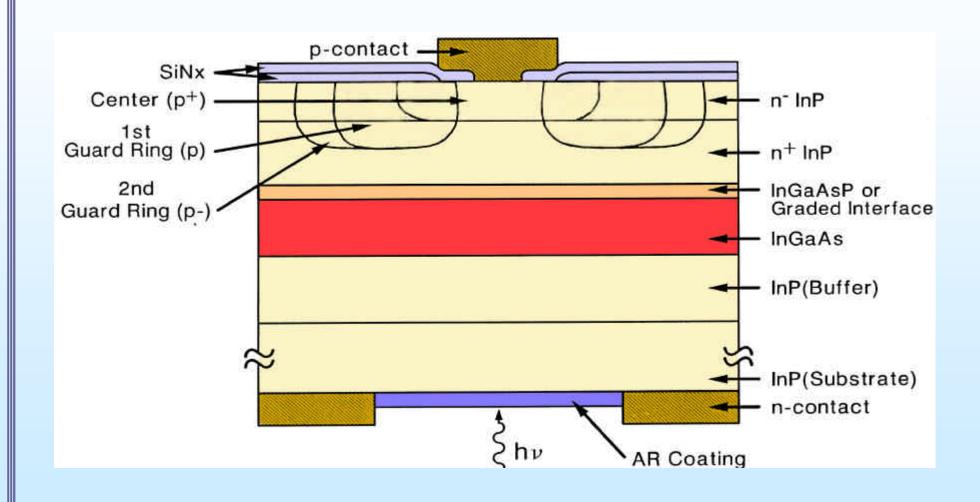




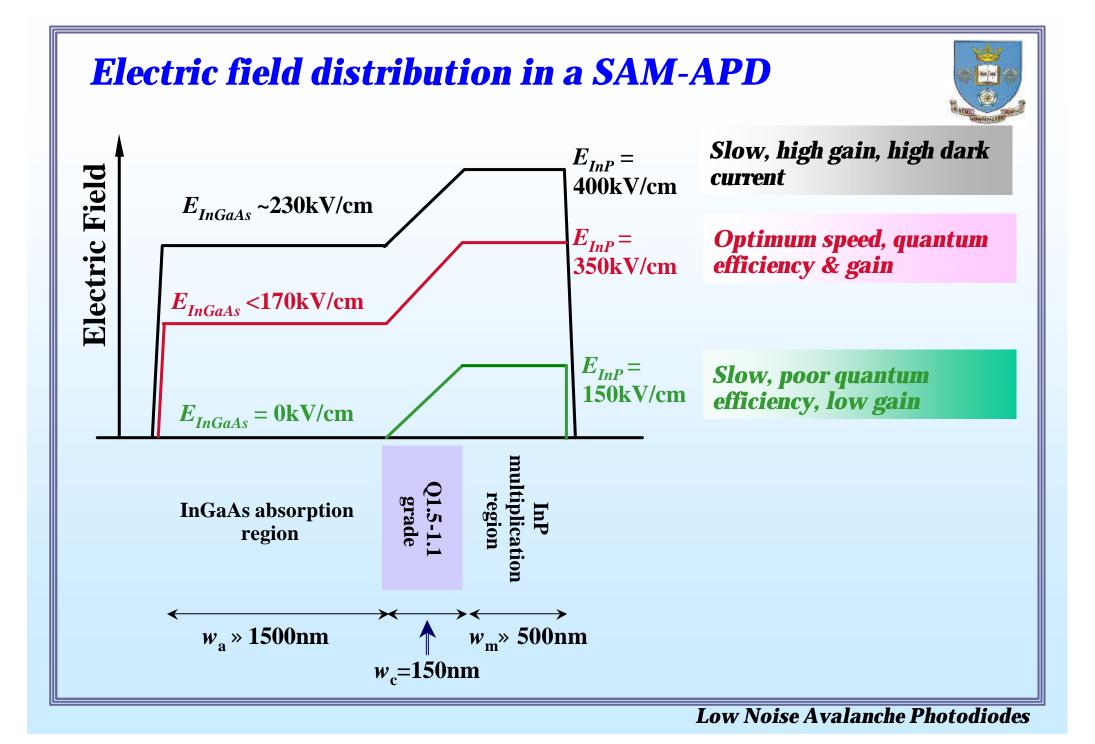
- APDs can have an extra -8dBm sensitivity c.f. p-i-n
- ⇒ APD advantage reduces as speed increases

InGaAs/InP SAM-APD



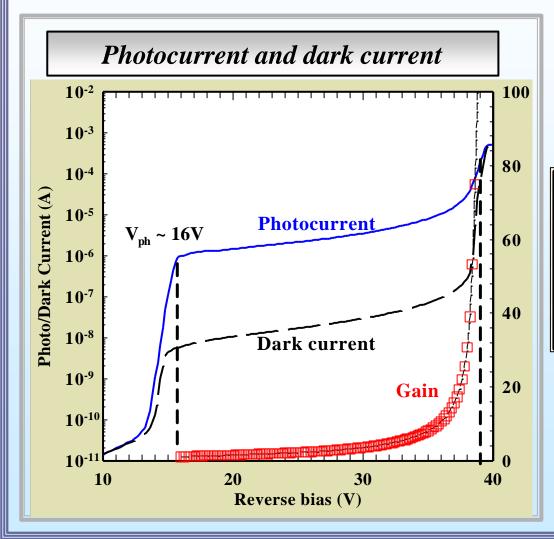


Low Noise Avalanche Photodiodes



SAM-APD Current-Voltage Characteristics





Punch-through ~ 16V Breakdown ~ 39V

Low dark current even at high multiplication gains (> 50) (< 70nA @ 0.9 V_{bd})

APD Vendors



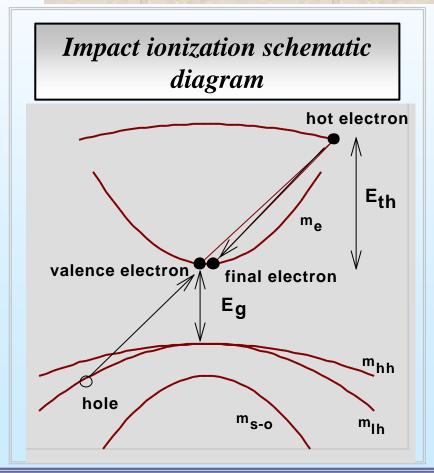
Vendors	Sensitivity	V _{BD} (V)	D V _{BD} (V/°C)	3DB bandwidth	Misc:
	(dBm)			(GHz)	
JDSU ERM577	-32	40-70		1.8	10nA@ V _{BD} -1.5
					8.5uA/uW@ V _{BD} -1.5
Mitsubishi	-33	35-75	0.12	1.9	
FU319SPA					
Agere	-34	45-70	0.07-0.14	2	
P173A					
Fujitsu	-34	50	0.12	2.5	
FRM5W232BS					
Nova Crystals	-	29-32	0.03		Fused InGaAs/Si
NVR251					
NEC 8501		40-80	0.1	3	7nA@0.9 V _{BD}
Alcatel 1914		30-75		3.5	70nA@0.9 V _{BD}
Sensors Unlimited	-24	25-42		6	5nA@ V _{BD} -1
SU-10ATR					
Mitsubishi	-25	20-40	0.05	7	
FU321SPA					
JDSU ERM578	-22	20-40		7	
NEC 4270	-25	16-32	0.02	8	Superlattice APD
					(k=0.4)

Ionization - Threshold energy



For ionization:

- 1) energy and momentum conservation
- 2) minimisation of energy

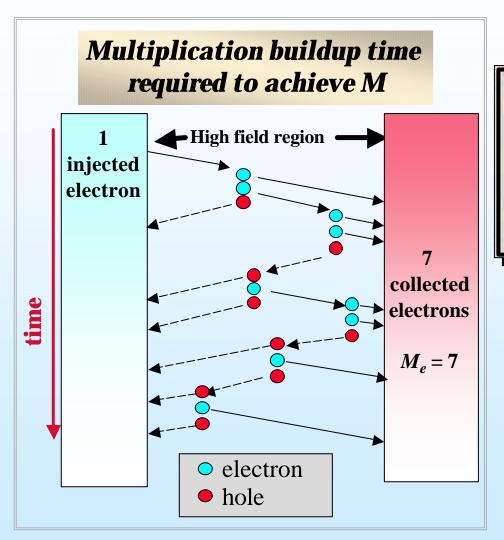


 E_g = Band gap E_{th} = Threshold energy
'minimum energy for ionization'

For parabolic bands and equal masses, E_{th} = 1.5 E_g

Multiplication process

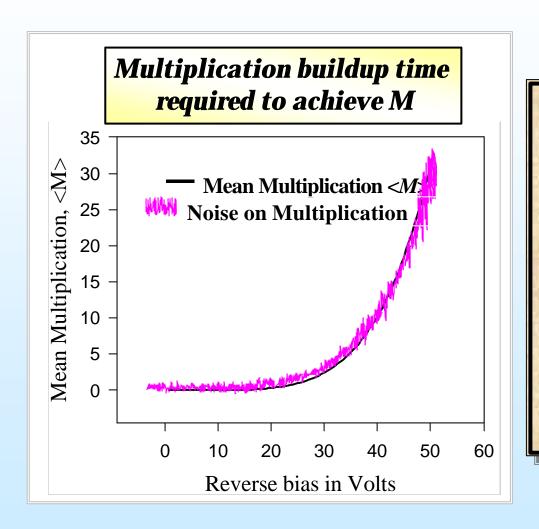




- Multiplication=
 current out/current in
- Avalanche takes time to build-up

Excess avalanche noise

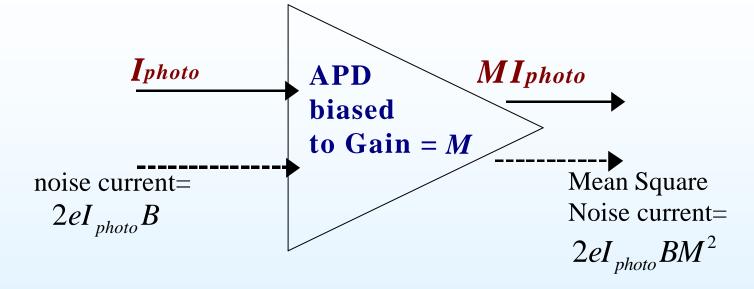




- An APD can give us gain.
- Unfortunately the avalanche noise can degrade the S/N ratio.
- An optimum value for < M> exists.

An ideal avalanche photodiode

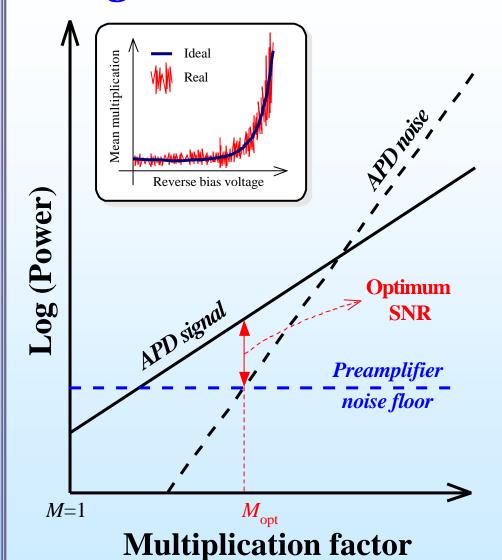




■ Non ideal APD $S = 2eI_{photo}BM^{2}F$ where F = excess noise factor For an ideal APD F = 1

Background



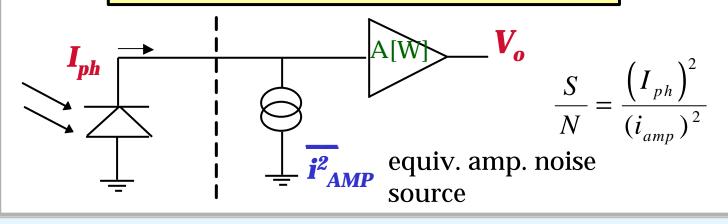


- ◆ APD's rely on internal gain to improve S/N ratio
- Impact ionization process
 stochastic > avalanche
 noise
- Excess avalanche noise limits APD's maximum useful gain, M
- In bulk structure, large b/a (or a/b) ratio required for low excess noise

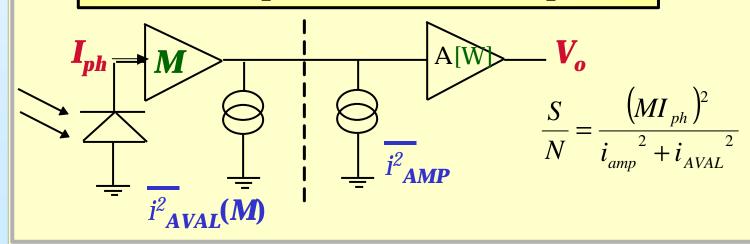
Photodetectors - S/N



p-i-n photodiode and amplifier



Avalanche photodiode and amplifier



McIntyre's avalanche noise theory (1966)



$$F(M) = M\left(1 + \frac{(1-k)}{k} \left(\frac{M-1}{M}\right)^{2}\right)$$

where
$$k = \frac{b}{a}$$

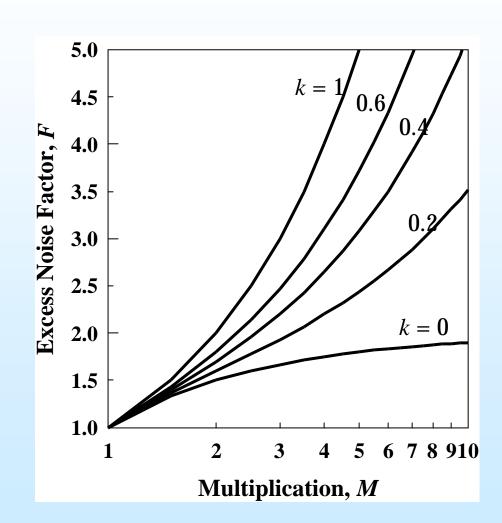
- \star a = electron probability of ionization per unit length [m⁻¹]
- ★ b = hole probability of ionization per unit length [m⁻¹]

Assumes:

- Multiplication process does not depend on carrier history.
- $\mathbf{a} \mathbf{k} = b/a$ is a constant

McIntyre's model for electron injection F v M_e





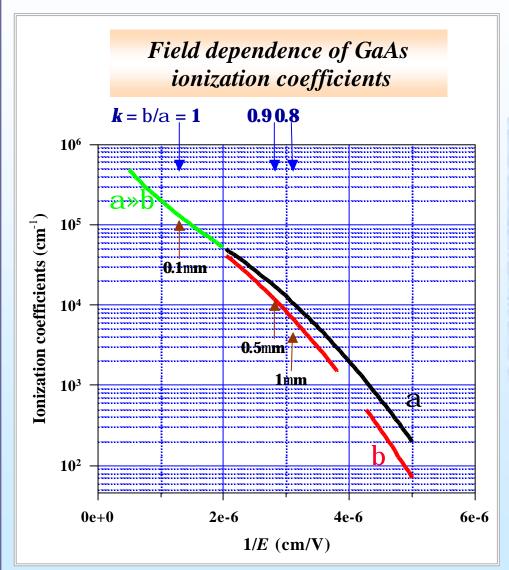
$$\mathbf{k} = b / a$$

The excess noise depends only on the ionization coefficient ratio (k) and the multiplication value.

Larger ionizing carrier type should initiate avalanche.

GaAs Ionization Coefficients

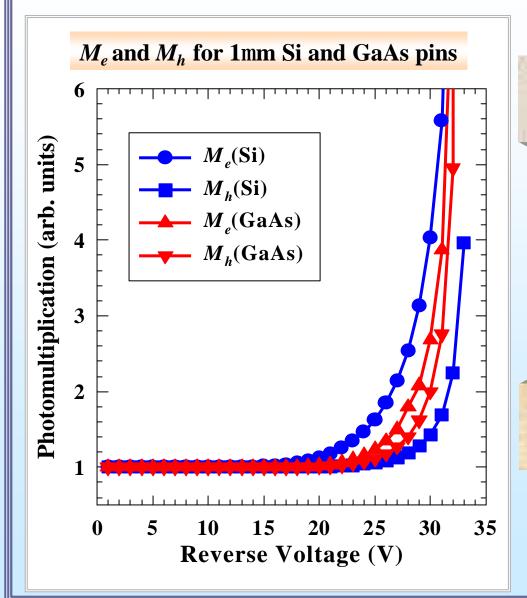




- * Most III-V semiconductors have 0.4 £ k £ 2.5
- * High excess noise expected, especially at higher electric fields when k® 1

Ionization Coefficients





Simple relationship between a, b and multiplication characteristics in pins.

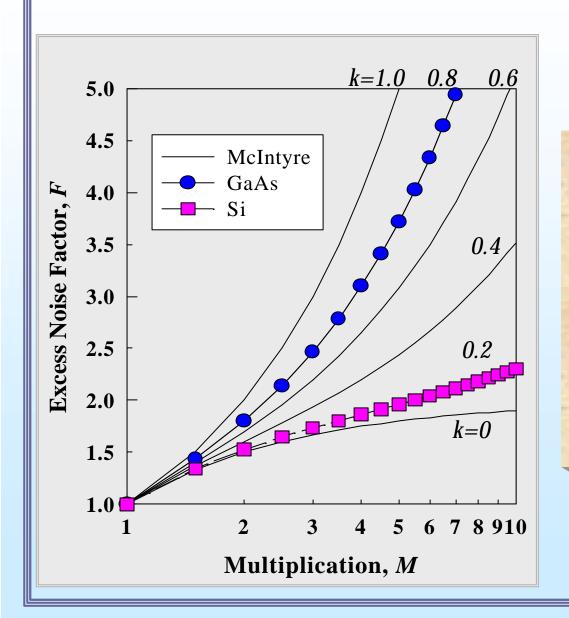
$$\alpha = \frac{1}{W} \left[\frac{M_e - 1}{M_e - M_h} \right] \ln \left(\frac{M_e}{M_h} \right)$$

$$\beta = \frac{1}{W} \left[\frac{M_h - 1}{M_h - M_e} \right] \ln \left(\frac{M_h}{M_e} \right)$$

GaAs & Si have similar V_{bd} but very different a/b ratios.

Excess noise in Si and GaAs, Me



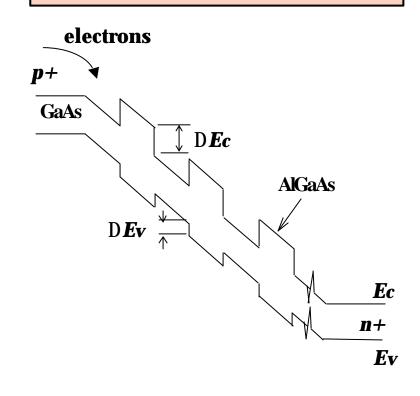


- In thick structures, the excess noise F is determined by k, the b/a ratio.
- Silicon has a small k compared to GaAs, hence low noise.

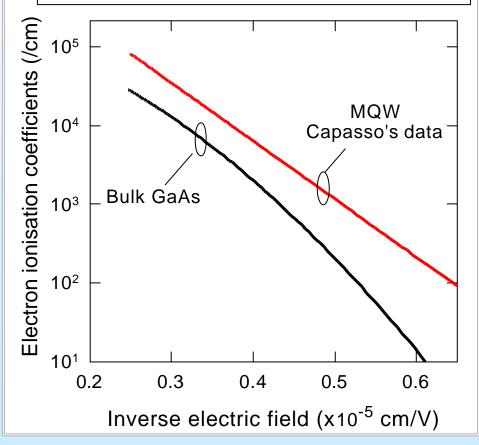
Enhancement of Ionization by MQW's



Chin et al. (1980) postulated that a large $\triangle Ec$ would enhance a

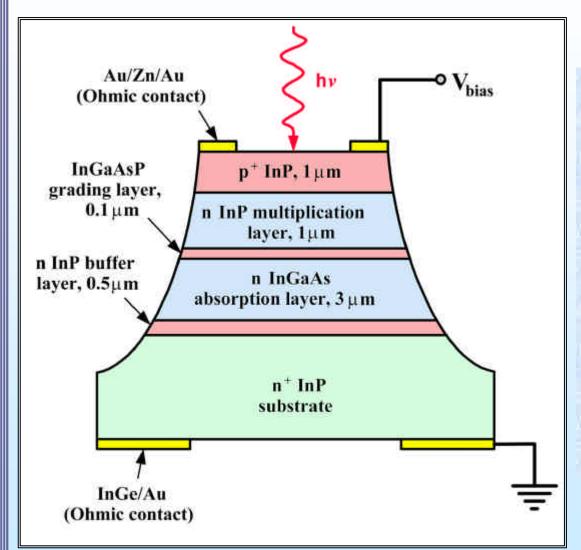


Capasso et al. (1982) reported a large enhancement in a ?in AlGaAs/GaAs MQW's



Schematic of a SAM-APD

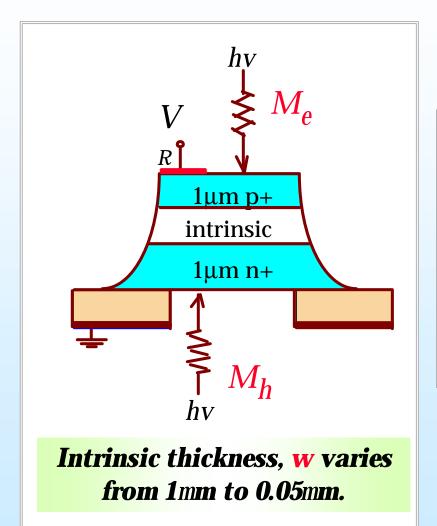




- Light is absorbed in thick InGaAs layer
- **○** Photogenerated holes impact ionize in InP
- Conventional designs involve thick multiplication layers, so that a/b ratio is small, to achieve low excess noise

p-i-n diode schematic



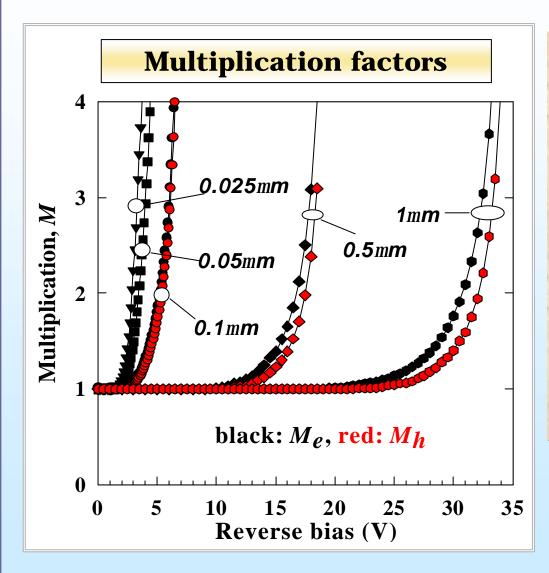


Pure M_e & M_h obtained by illuminating thick p+ & n+ layers with short wavelength illumination.

 n^+ -i- p^+ s also grown to obtain M_h more easily.

Multiplication from GaAs p⁺-i-n⁺s

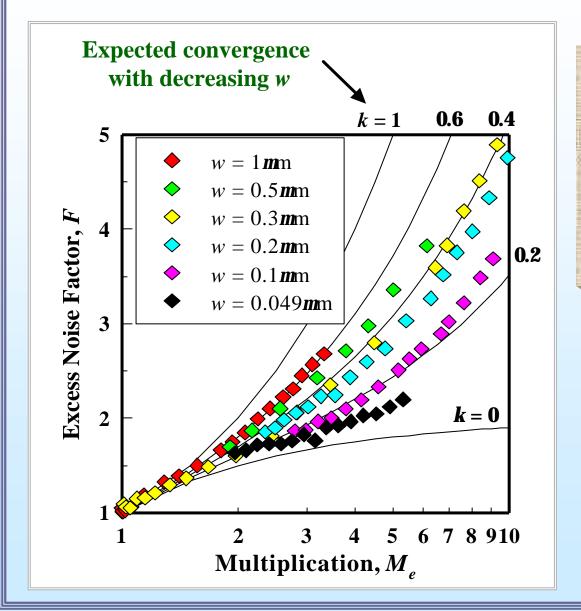




- M_e and M_h were measured in different thickness p⁺-i-n⁺s.
- Lock-in techniques allow M_e and M_h to be determined in the presence of large dark currents.
- $M_e \gg M_h$ as 'w' decreases, suggesting that a \gg b

Excess noise in GaAs p⁺-i-n⁺s



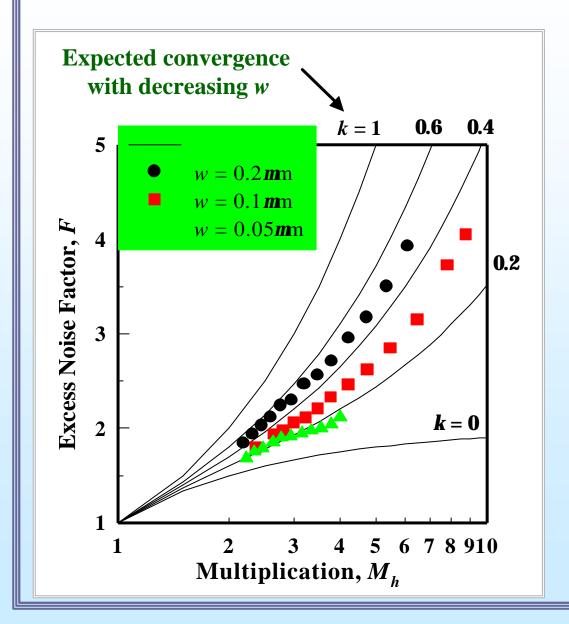


Electron initiated noise measurements showed unexpected and significant noise reduction as w became smaller

The excess noise decreases as w decreases, instead of increasing as k® 1

Excess noise in GaAs n⁺-i-p⁺s





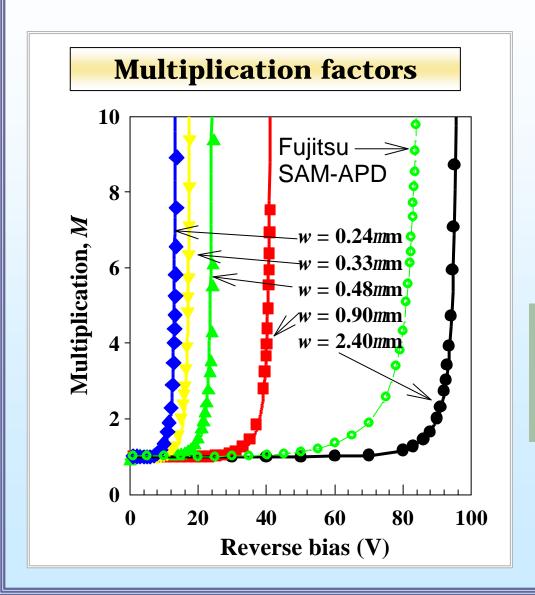
The excess noise *decreases* as w decreases, instead of increasing according to k.

Hole initiated noise measurements also showed unexpected and significant noise reduction as w became smaller

Behavior cannot be explained by McIntyre theory

Multiplication characteristics in InP



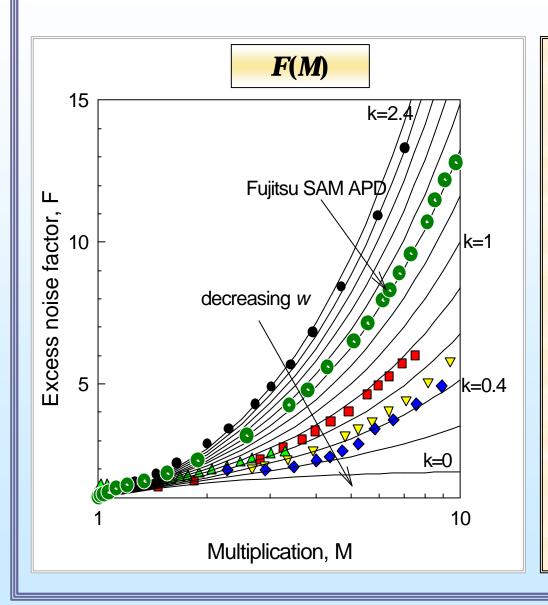


Measured M_e (symbols)

Calculated M_e (solid lines) using bulk ionization coefficients

Excess noise factor in InP

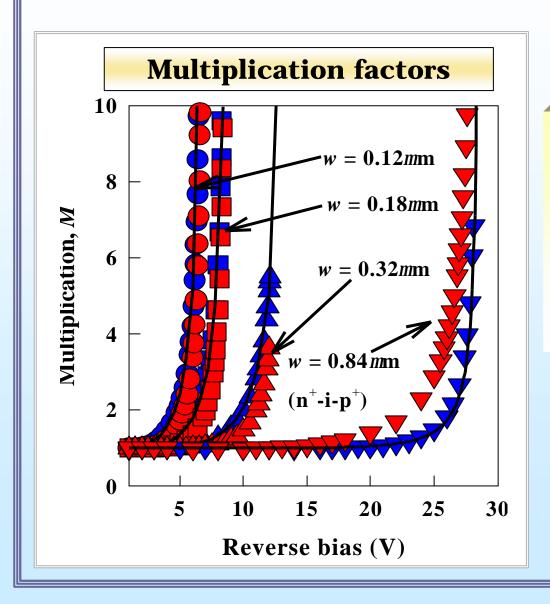




- ⇒ Same symbols as before
- Noise measured using wrong (electron) carrier type
- Fujitsu SAM-APD gives $b/a = 1.4 \setminus k_{eff} = 0.7$
- Structure with w = 0.24 gives k_{eff} = 0.4 much better than SAM-APD with hole multiplication

Multiplication characteristics in Silicon



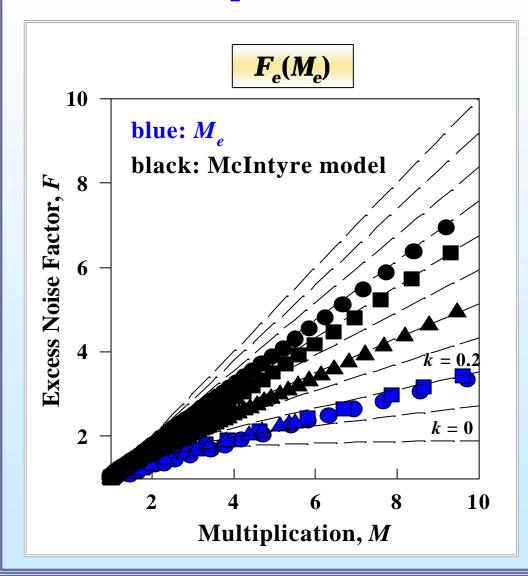


Measurement of M_h and M_{mix} on 0.84mm n⁺-i-p⁺

Measurement of M_e and M_{mix} on 0.32, 0.18, 0.12mm p⁺-i-n⁺s

Local noise model prediction vs. experiment in submicron Si p⁺-i-n⁺s





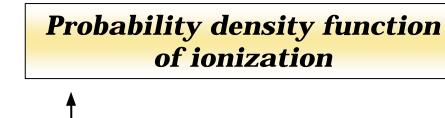
- Local field noise model gives increasing excess noise from k = 0.4-0.7 as w decreases from 0.32-0.12mm.
- Experiment shows that $F(M_e)$ however is virtually constant at $k \gg 0.2$.

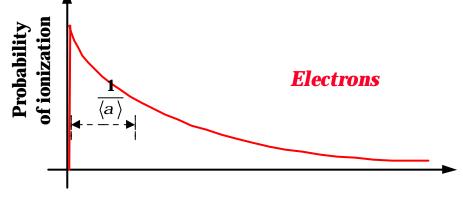


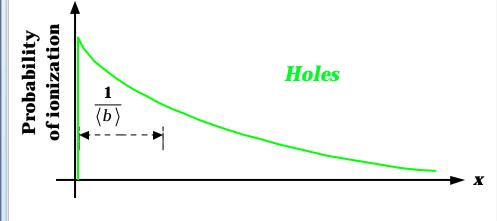
Modeling of thin APD behavior

McIntyre Noise Model







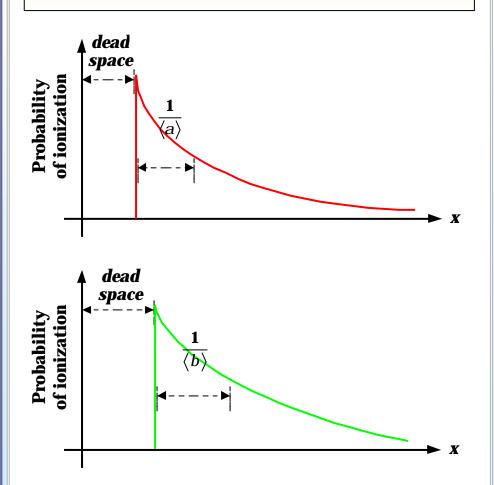


- ★ McIntyre's noise model
 assumes that a carrier's
 ionization probability is
 independent of distance ¾
 probability density function
 (PDF) is exponential
- **★** This assumption leads to the McIntyre expression for excess noise factor
- **★ Avalanche noise depends on** the *b /a* ratio

Dead Space Models



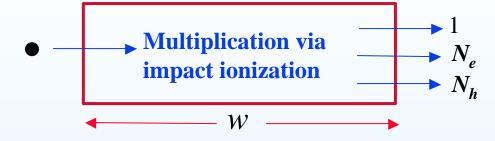
Probability density function of ionization



- ★ More realistic picture of ionization probability shows significant dead space at high electric field
- ★ Presence of dead space reduces CoV ¾ makes multiplication more deterministic ¾ less noisy
- **★ A significant dead space**reduces the importance of
 the b/a ratio & the carrier
 type initiating
 multiplication

Monte Carlo Estimation of F





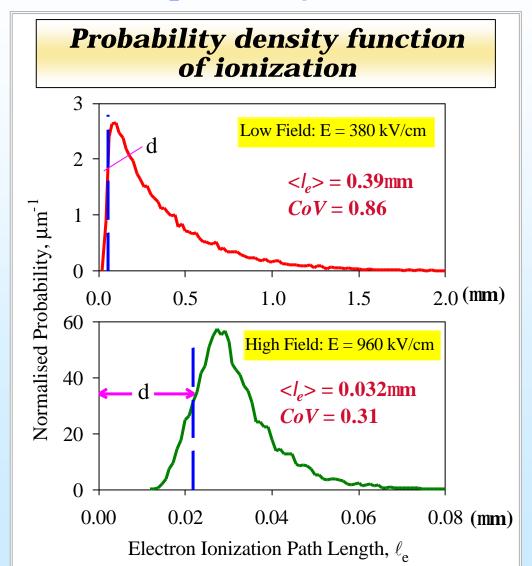
$$M_{trial} = 1 + N_e + N_h$$

$$< M >= \frac{(M_1 + M_2 + ... + M_{N-1} + M_{N-2})}{N}$$

$$< M^2 >= \frac{(M_1^2 + M_2^2 + ... + M_{N-1}^2 + M_{N-2}^2)}{N}$$
Excess Noise Factor, $F = \frac{< M^2 >}{< M >^2}$

Probability distribution of electron ionization path lengths $(\langle M \rangle = 5)$



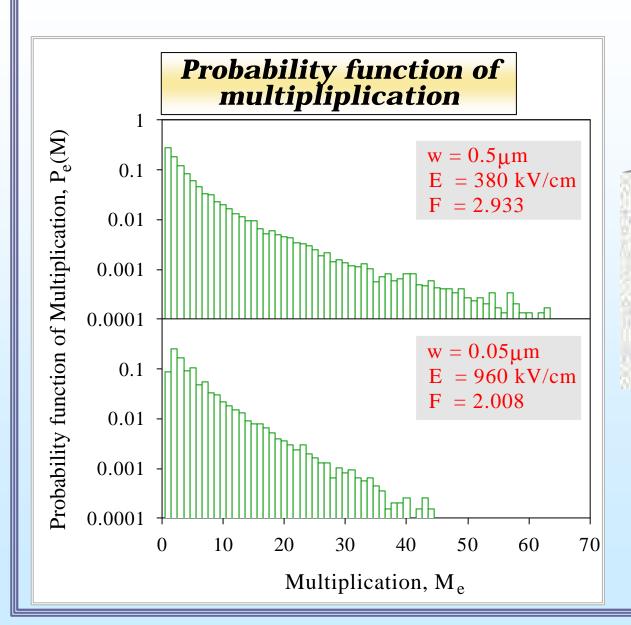


- At low fields ⇒ relatively small dead space & low ionization probability
- At high fields ⇒ relatively large dead space & higher ionization probability ⇒ narrow ionization probability distribution.

 $CoV = stand. dev. in <math>l_e / < l_e >$

Distribution of Multiplication for $\langle M \rangle = 5$

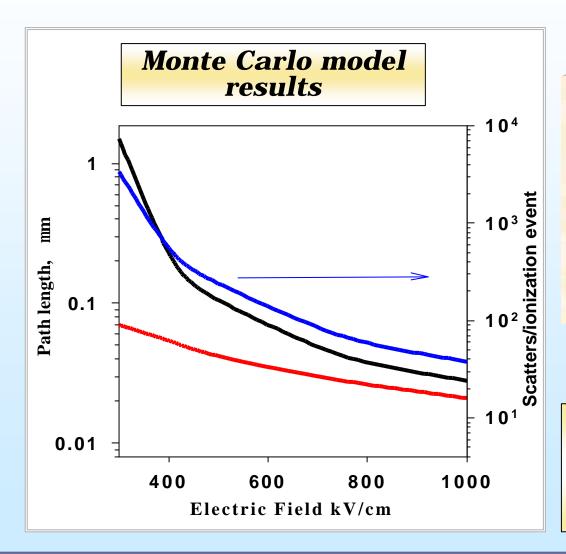




 There are more high order multiplication events at lower electric fields, giving rise to more noise

Typical path lengths as a function of electric field



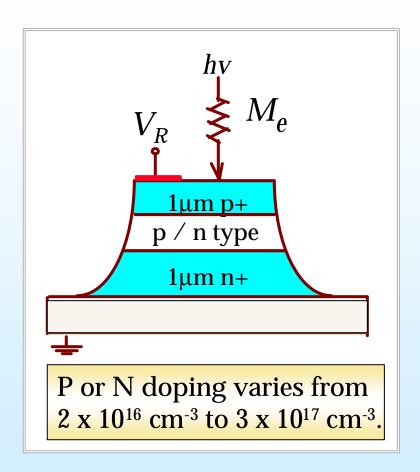


- Scattering becomes
 less important as the
 electric field increases
- Ionization tends towards ballistic ideal, i.e. like PMT

Scatters per ionization event
 Ballistic dead space
 d = 2.1eV/qE
 Mean ionization path length < l_e>

Excess noise in p⁺-n diodes



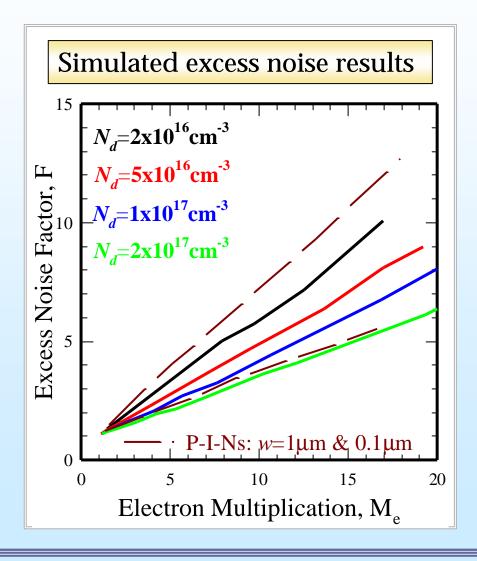


- → Reducing w in p⁺-i-n⁺s reduces excess noise.
- → How does increasing doping i.e. electric field gradient affect noise?

Simulated excess noise in P⁺N junctions

(... primary electrons are injected into the HIGH field)





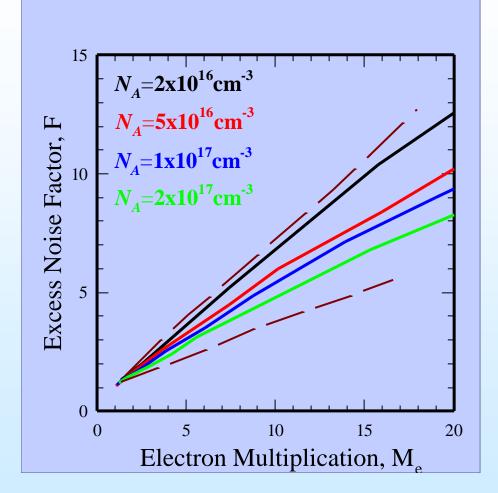
The excess noise is reduced by increasing the doping

For the same total depletion thickness, F(P+N) < F(P+I-N+)

Simulated excess noise in PN⁺ junctions



(... primary electrons are injected into the LOW field)

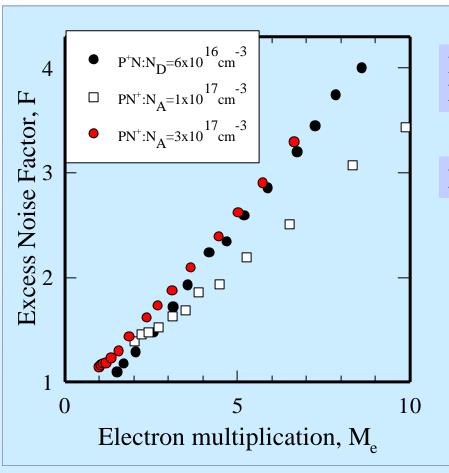


The excess noise is again reduced by increased doping

For the same doping magnitude, $F(PN^+)$ is slightly greater than $F(P^+N)$

Experimental results with electron injection





Increasing the doping in the PN⁺ devices reduces noise

Experiment corroborates theory

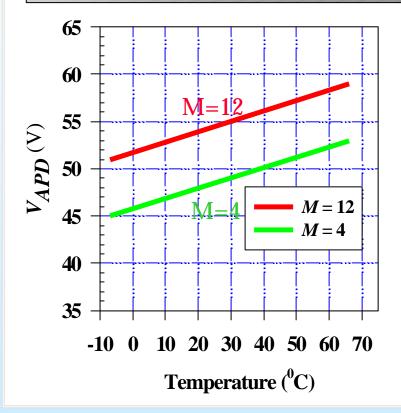


Effect of temperature variation on APD performance

Temperature dependence of avalanche multiplication





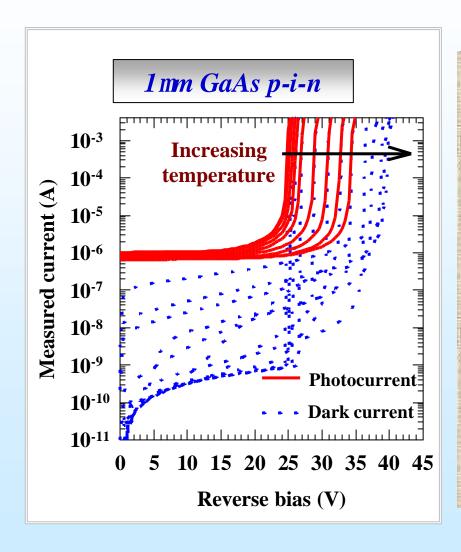


- APD multiplication is very temperature sensitive
 Not a problem when input signal is large - BER increases when at the limit of sensitivity
- Breakdown variation is ~
 0.06-0.2V/°C

Active circuit required to vary bias to ensure constant multiplication

Temperature dependent I-V for 1 mm GaAs

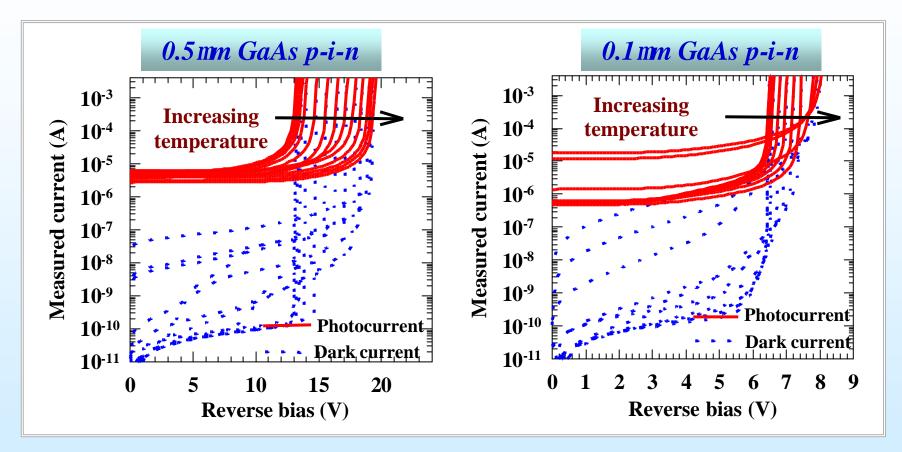




- Photocurrent, dark current and breakdown measured on different thickness GaAs p-i-n diodes, from 20K-500K
- Sharp V_{bd} observed at all temperatures.
- Dark currents increase with temperature
- Avalanche multiplication reduces with increasing temperature

Temperature dependent I-V for 0.5 mm & 0.1 mm GaAs

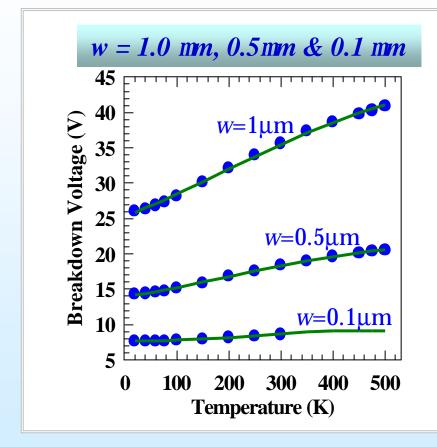


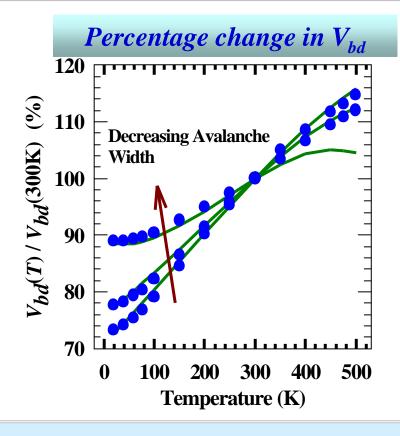


- Similar behavior seen in thinner avalanche width structures
- Thinner devices are less affected by changes in temperature

Change in V_{bd} with Temperature





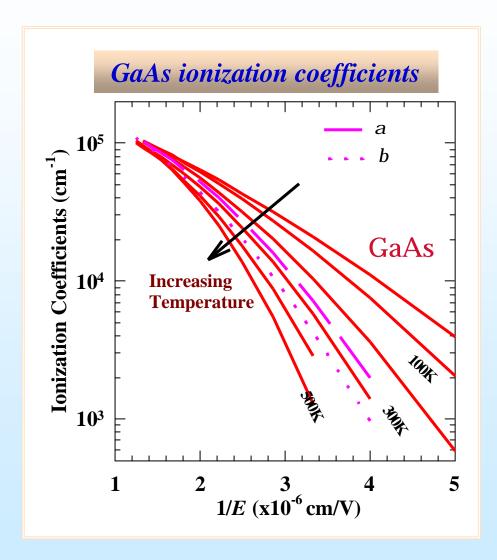


The breakdown change is more significant in thicker structures

Temperature coefficient decreases from 0.032V/°C to 0.004V/°C

Temperature dependent ionization coefficients

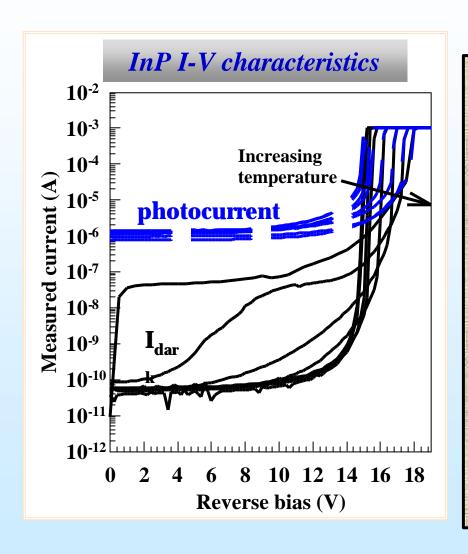




- Ionization coefficients derived from multiplication data
- Ionization coefficients decrease with increasing temperature
- The change is much larger at lower electric fields
- Thinner avalanche widths operate at higher electricfields
- Phonon scattering relatively less important at higher electric fields

Temperature dependent I-V in thin InP

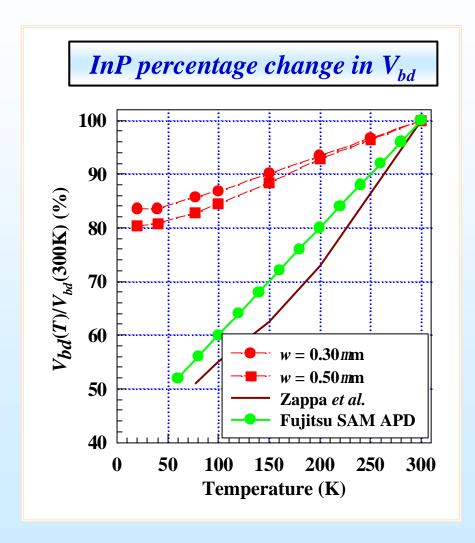




- Similar measurements on w = 0.3 mm InP p-i-n from 20K-300K
- Low dark current and
- Sharp breakdown observed
- V_{bd} decreases as temperature decreases
- Similar results observed in structure with w =
 0.5 mm.

InP temperature coefficient of V_{bd}





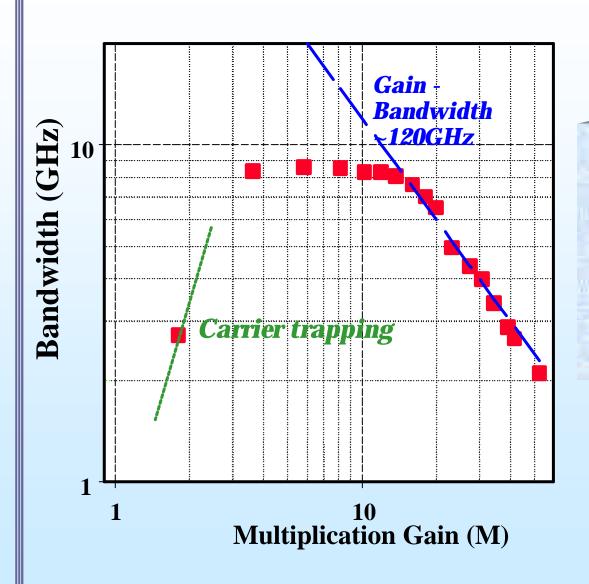
- → Lower temperature coefficient of breakdown voltage, h_o as w decreases.
- → Zappa et al (IPRM'96): h_o ~ 0.225V/°C
- → Fujitsu SAM-APD: h_o ~ 0.09
 0.15V/°C
- \Rightarrow $w = 0.3 \text{mm: } h_o \sim 0.012 \text{V/}{}^{\circ}C$
- \rightarrow **w** = **0.5**m**m**: $h_0 \sim 0.02V/{}^{\circ}C$



Effect of thin avalanching widths on APD speed

Gain-bandwidth Characteristics



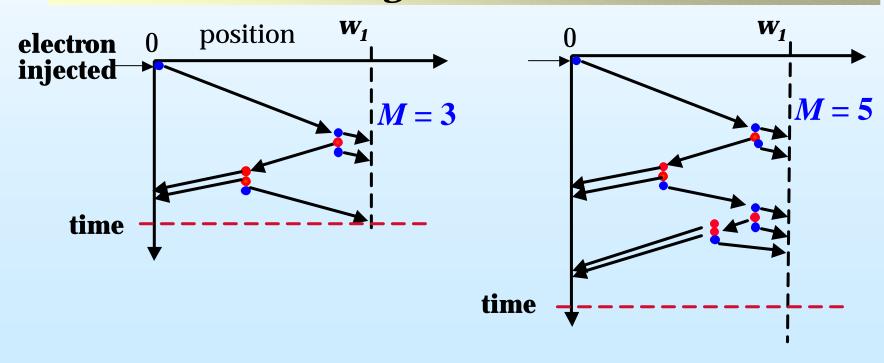


- Bandwidth decreases at low multiplication carrier trapping
- Bandwidth decreases at high multiplication multiple transits

APD speed limitations-multiplication build-up time

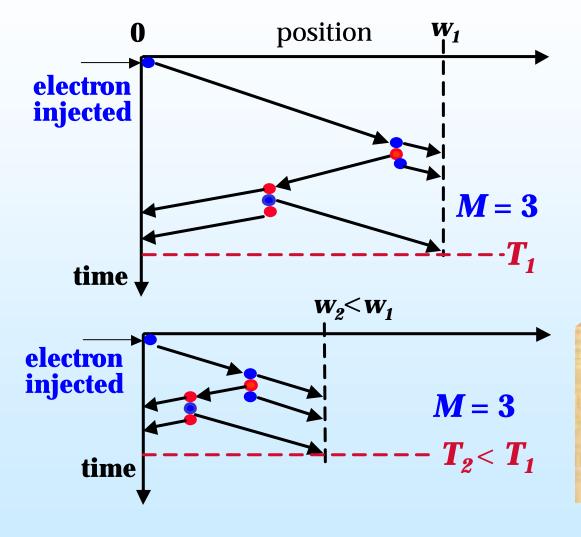


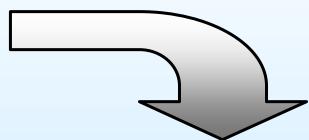
- APD is slow c.f. p-i-n diodes due to multiple transits required for high gains
- Difficult to achieve 10 Gb/s operation with thick avalanching structures



Thin avalanche region multiplication build-up time





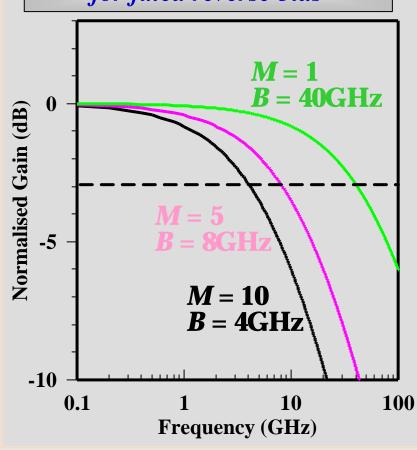


Decreasing w results in shorter transit times higher speed

APD limitations - frequency response





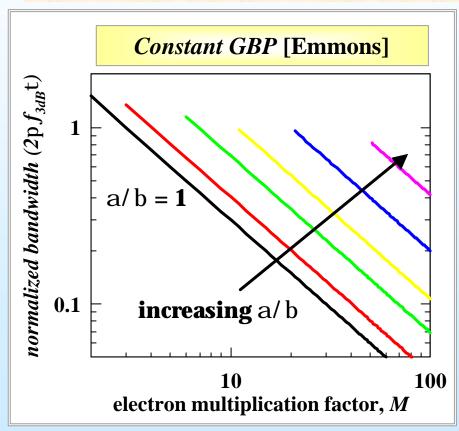


- APD frequency response approximates a 1st order system
- ◆ Figure of merit Gain bandwidth product (GBP)
- Motivation of thin avalanche regions < 1mm to increase GBP</p>

Multiplication-limited bandwidth



* Factors affecting APD speed: Carrier transit time, RC time constant, carrier diffusion and multiplication buildup time (f_{3dB})

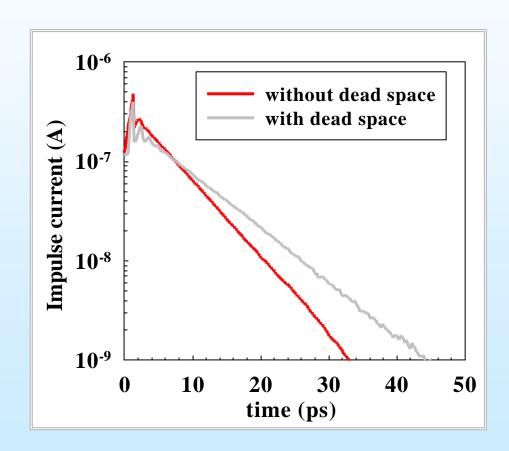


- Conventional (Emmons') model [Emmons, 1967]
 - ♦ Negligible dead space \Rightarrow d = 0
 - **♦** Constant carrier speed $\Rightarrow v = v_{sat}$ (saturated drift velocity)
 - Constant gain-bandwidth product, GBP
 - GBP scales with t, carrier transit time

Effect of dead space on speed



* Comparison of d = 0 cf. nonlocal model with d 1 0



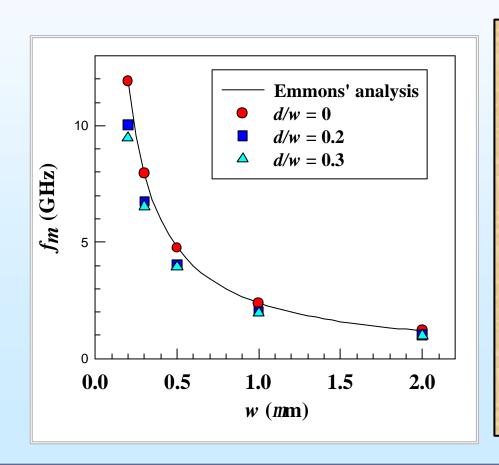
$$\Re v = v_{sat}$$

- * Avalanching region width of w = 0.1 mm
- *M = 12.5
- ** d û, avalanche current impulse response decays more slowly ⇒ lower f_{3dB} [Hayat and Saleh, 1992]

Effects of dead space



- f_m of APDs with avalanche width of w, a = b and
 <M> = 20
- Compare d/w = 0, 0.2 and 0.3

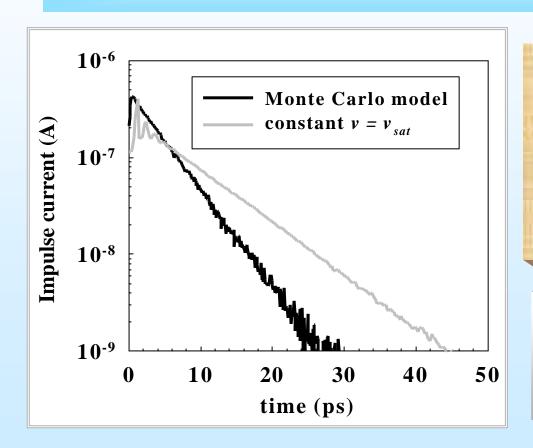


- Obtain f_m by Fourier transforming the current impulse response
- f_m obtained agrees with Emmons' prediction for d/w = 0
- As d/w increases, f_m falls below the predicted values
- ♦ d/w is larger in thin APDs
 ⇒ absolute decrease in f_m is larger in thin APDs

Carrier speed assumptions



- ***** Monte Carlo model cf. constant $v = v_{sat}$ model
- Same dead space, d

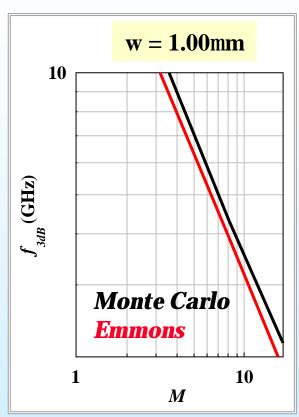


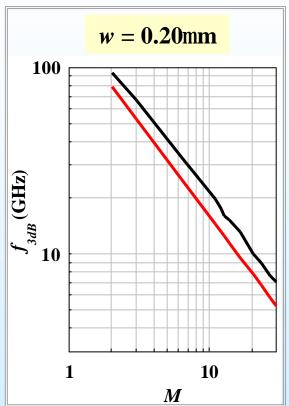
- * w = 0.1mm, M = 12.5
- *** Enhanced speed** in MC model leads to faster decay of current impulse response ⇒ higher f_{3dB}
- ★ Dead space and enhanced speed effects compete!

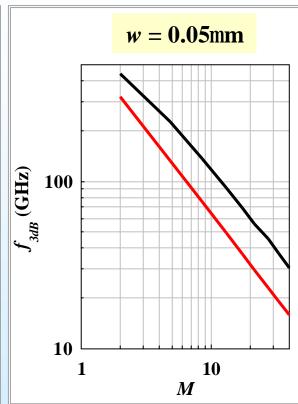
 [Hambleton et al, 2002]

Simulation result comparisons





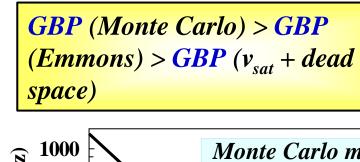


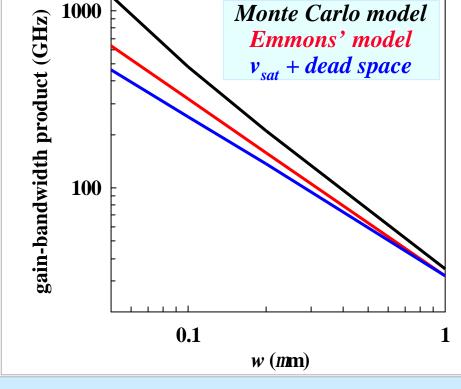


- Constant GBP
- f_{3dB} (Monte Carlo) > f_{3dB} (Emmons) for all w and all M
- Enhanced carrier speed dominates dead space

GBP comparison







- ★ Monte Carlo ↑ more rapidly as w ↓ than Emmons and (v_{sat} + dead space)
- * GBP enhancement ↑
 as w ↓
- ♦ Worst case is using (v_{sat} + dead space)

Published experimental f_{3dB}



- Lenox et al. (PTL 1999) measured f_{3dB} of InAlAs RCE APDs
 - $\odot W = 400 \text{ nm } GBW = 130 \text{ GHz}$
- \odot $GBP_{200nm} > 2$ GBP_{400nm}
 - \Rightarrow **Emmons' model** predicts $GBP_{200nm} = 2$ GBP_{400nm}
 - > But larger d/w in w = 200nm device slows frequency response
 - Suggests V_{200nm} > V_{400nm}

Conclusions



- **★** The excess noise decreases as the avalanche width decreases below 1 mm, in disagreement with the theory of McIntyre
- **★** The low noise results from a more deterministic impact ionization process at high fields as *dead space* becomes more important
- **★** The carrier type initiating the multiplication becomes unimportant at high fields
- **★** Thin avalanching regions should be less temperature sensitive
- ★ Thin avalanching regions should be capable of high speed operation
- **★ 40 Gb/s APDs highly probable**

Acknowledgements



Graham Rees, Peter Robson, Richard Tozer, Bob Grey, Mark Hopkinson, Geoff Hill, J.S. Roberts

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IEEE LEOS, EPSRC, DERA, EU

UV Detection

Visible-blind

Solar-blind

200

300

400

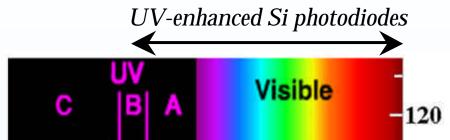
500

Wavelength (nm)

600

700





- Applications requiring **UV** detection
- **►** Atmospheric UV remote sensing
- Combustion control
- corona discharge on HV lines

Motivation

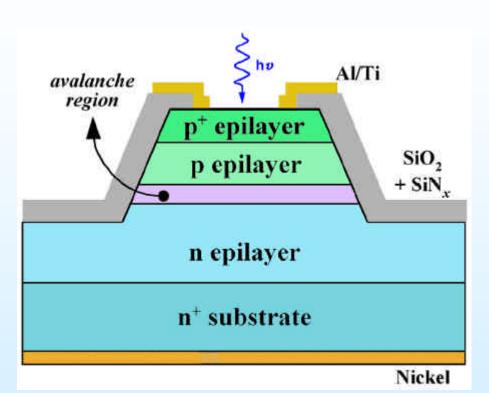


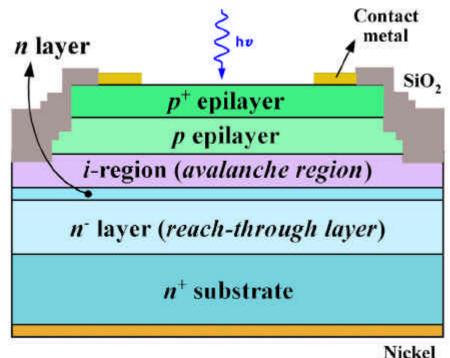
Why SiC for UV APDs?

- Wide bandgap (3.25eV for 4H-SiC)
 - ⇒ excellent for UV detection
 - ⇒ very low dark current
 - ⇒ high temperature operation
- Large b/a ratio in 4H-SiC
 - ⇒ desirable for thick APD structures
 - ⇒ performance in thin structures unknown

4H-SiC Device Structures



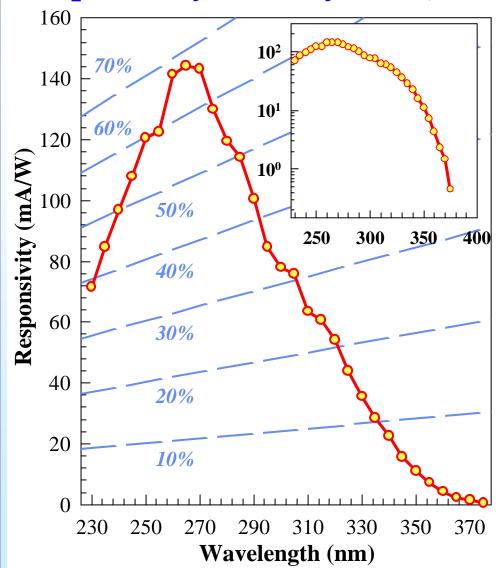




- 2° +ve bevel edge & multistep junction extension termination
- Square mesas with areas: 50 ′ 50 ~ 210 ′ 210 mm²
- Passivated with SiO₂ & SiN_x
- Al/Ti top contact with optical access

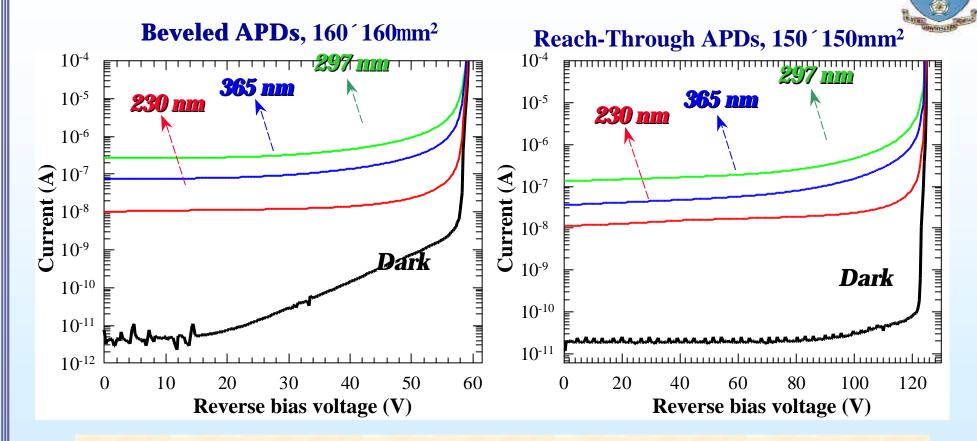
Responsivity at Unity Gain, Beveled APDs





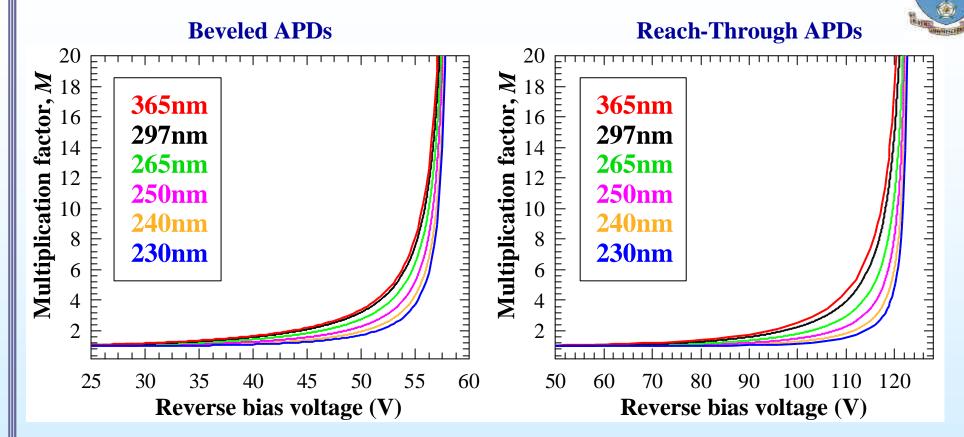
- Similar to typical 6H-SiC photodiodes
- Responsivity cutoff at
 ~380 nm visible-blind
- Peak responsivity of 144
 mA/W at 265 nm p
 quantum efficiency of ~
 67%

Reverse IV Characteristics



- Avalanche breakdown is sharp & well-defined at $V_{bd} = 58.5$ V & 124.0V
- Carriers injected with 230 ~ 365 nm light to initiate multiplication
- I_{ph} is 1 ~ 3 orders of magnitude > I_{dark}
- AC measurements corroborate DC results

Multiplication Characteristics



- M of > 200 measured
- M at various 1 more disparate for thicker APD structure
- Smaller M from shorter l

$$P M_b > M_e P b > a$$

Excess Avalanche Noise Characteristics





30

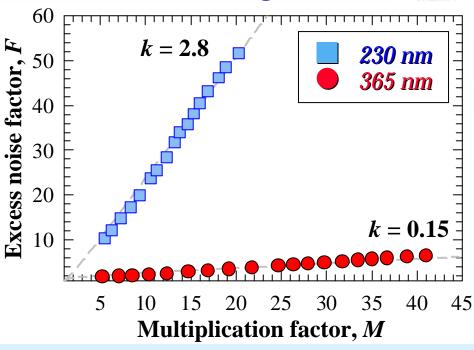
Multiplication factor, M

30 Excess noise factor, Fk = 0.8230 nm 25 365 nm 20 15 10

20

10

Reach-Through APDs



• Excess noise measured for M > 40**P** good quality of APDs, very stable avalanche multiplication

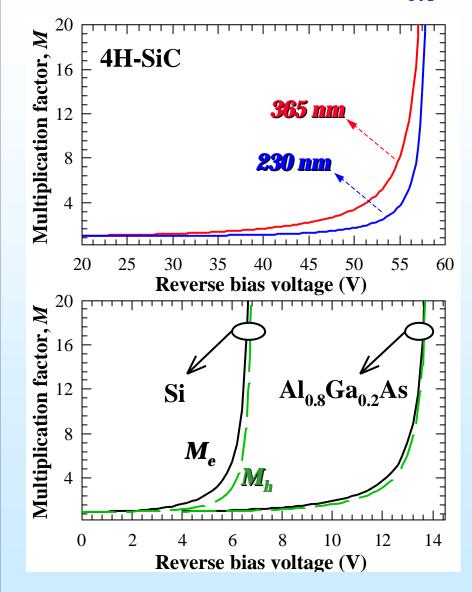
40

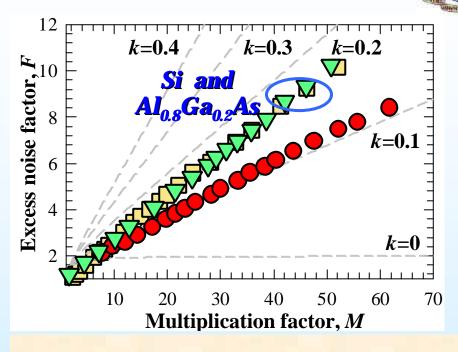
• Very low excess noise of k = 0.1 & 0.15 measured with 365 nm light

50

• Excess noise from electron injection (230 nm) gave k = 0.8 & 2.8

Comparison with Si & Al_{0.8}Ga_{0.2}As





- V_{bd} of 4H-SiC is 10´ & 5´ of Si & Al_{0.8}Ga_{0.2}As respectively
- M_e & M_h closer for Si, $Al_{0.8}Ga_{0.2}As$
- 4H-SiC > lowest excess noise in a
 w = 0.1 mm structure

Conclusions



- **4H-SiC APD's exhibit good visible-blind performance**
- **Photomultiplication characteristics**
 - ▶ Large M in excess of 200 measured
 - \triangleright show unambiguously that $\beta > \alpha$
 - $\triangleright \beta/\alpha$ ratio remains large in short devices
- Very low excess noise of k = 0.1 achieved with mainly holes-initiated multiplication
- 4H-SiC is a suitable material for high-gain, low noise UV avalanche photodiodes

Al_{0.8}Ga_{0.2}As: A Very Low Excess Noise Multiplication Medium for GaAs-based APDs



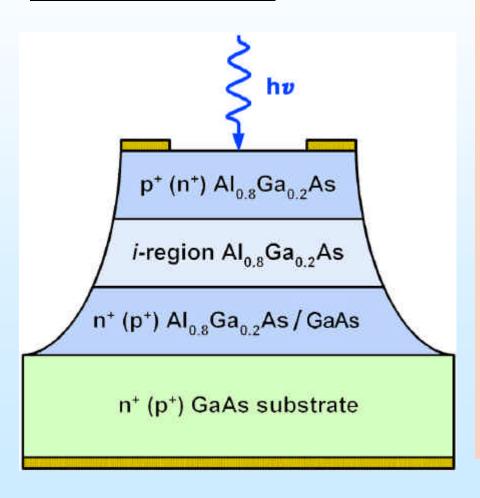
Motivation

- Al_xGa_{1-x}As material system is widely used in HBTs and IMPATTs
- Use in telecom wavelength APDs limited by lack of lattice-matched material that absorbs at long wavelength
- GaInAsN has recently been demonstrated
 - absorbs long wavelength
 - ⇒ lattice-matched to Al_xGa_{1-x}As
- GaAs-based APDs is possible and may require Al_xGa_{1-x}As multiplication region for optimum performance

Al_{0.8}Ga_{0.2}As: A Very Low Excess Noise Multiplication Medium for GaAs-based APDs



Device structures

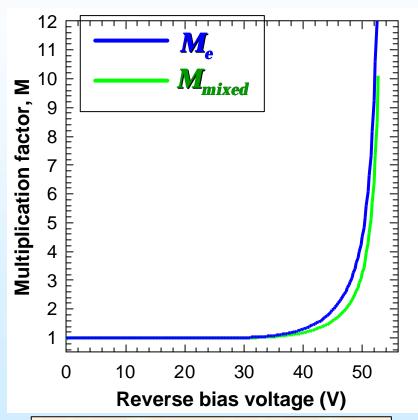


- Homojunction p-i-n/n-i-p grown by conventional MBE with w = 1 mm
- 1 heterojunction p-i-n with w=0.8 mm to obtain $M_e \& M_h$ from same diode
- Optical access window fabricated by wet etching
- Pure carrier injection obtained with 442nm & 633nm light
- 542nm light used to produce mixed carrier injection

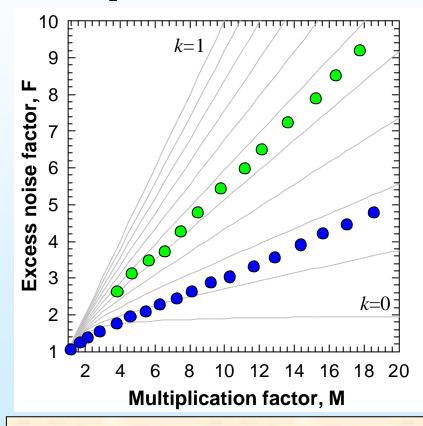
$Al_{0.8}Ga_{0.2}As: A Very Low Excess Noise Multiplication Medium for GaAs-based APDs$



Avalanche excess noise of homojunction p-i-n diodes



Lower M for mixed carrier injection P a > b

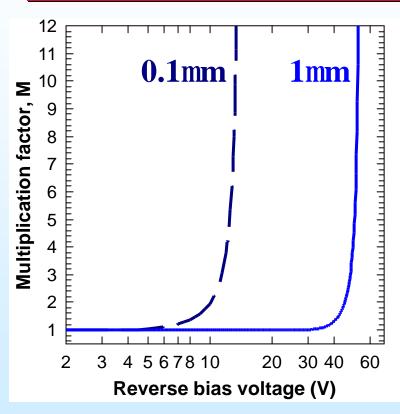


- k~ 0.19 for electron multiplication
- Larger F for mixed carrier injection

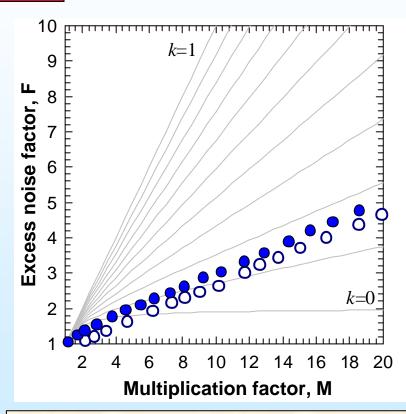
$Al_{0.8}Ga_{0.2}As: A \ Very \ Low \ Excess \ Noise \ Multiplication \ Medium for \ GaAs-based \ APDs$



Avalanche excess noise of thin diodes



• V_{bd} with decreasing w

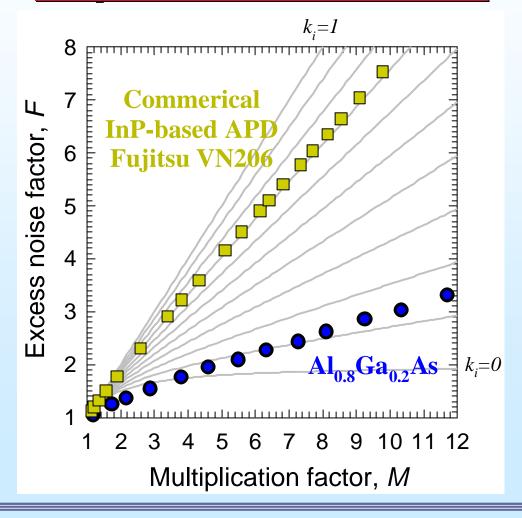


 Comparable excess noise for bulk and thin diodes

Al_{0.8}Ga_{0.2}As: A Very Low Excess Noise Multiplication Medium for GaAs-based APDs



Comparison with InP-based APDs

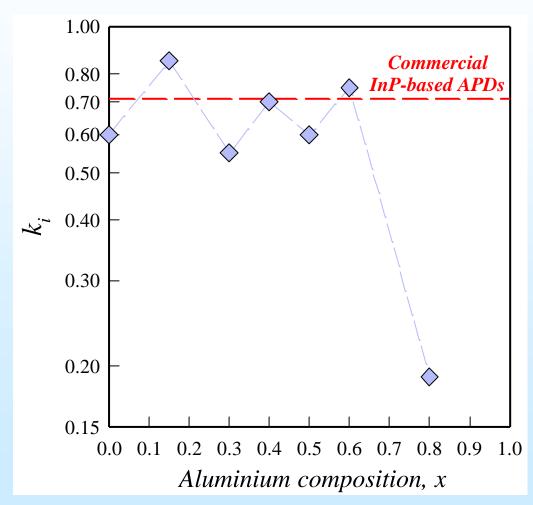


- Commercial InP-based APD give excess noise of k_i =~0.7 with hole initiated multiplication
- Much lower excess noise can be obtained with Al_{0.8}Ga_{0.2}As as avalanche medium

Al_{0.8}Ga_{0.2}As: A Very Low Excess Noise Multiplication Medium for GaAs-based APDs



Comparison with lower aluminium Al, Ga_{1-x}As

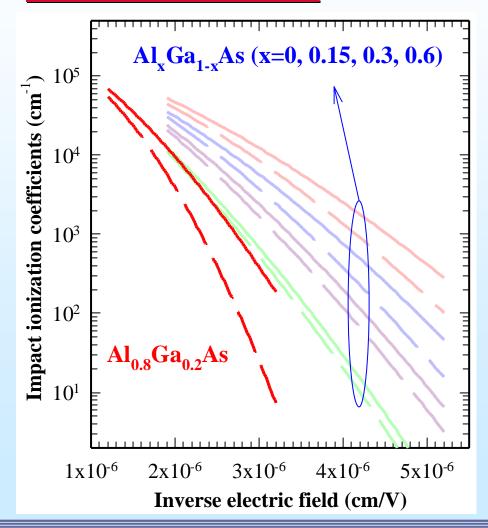


- Al_xGa_{1-x}As (x £ 0.6) has large avalanche excess noise
- Excess noise of Al_{0.8}Ga_{0.2}As is much lower
- Al_{0.8}Ga_{0.2}As also has lower excess noise than a commercial InP-based APD
- At M=10, excess noise of Al_{0.8}Ga_{0.2}As is at least 2 times lower

$Al_{0.8}Ga_{0.2}As: A Very Low Excess Noise Multiplication Medium for GaAs-based APDs$



Ionization coefficients



- Large a/b ratio as compared to Al_xGa_{1-x}As of lower x
 - **P** Lower excess noise
- b/a ratio of InP is small
 Higher excess noise

Al_{0.8}Ga_{0.2}As: A Very Low Excess Noise Multiplication Medium for GaAs-based APDs

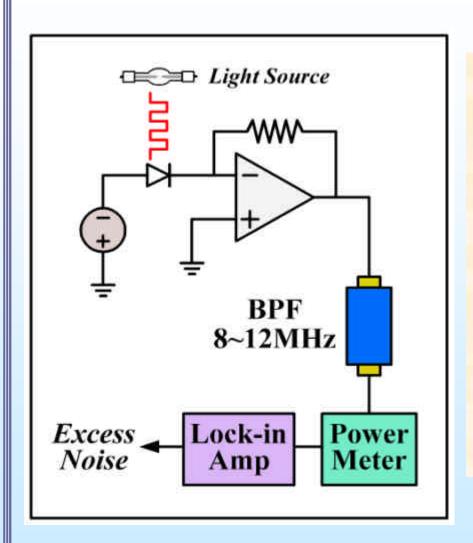


Conclusions

- Bulk $Al_{0.8}Ga_{0.2}As$ diodes give lower excess noise than $Al_xGa_{1-x}As$ (x £ 0.6) or InP
- Consequence of the larger a/b ratio in $Al_{0.8}Ga_{0.2}As$
- Low noise APDs may be achievable on GaAs substrates using Al_{0.8}Ga_{0.2}As as the gain medium

APD noise measurement system





Excess noise

- 10 MHz center frequency
- **ENBW of 4 MHz**
- AC technique with modulated light source
- F from M & noise power
- Shot noise of Si p-i-n as reference